

RC filters and signal conditioning

MEC100x-Lectures 4

Energy, Power and Intelligent Control

School of Electronics, Electrical Engineering and Computer Science

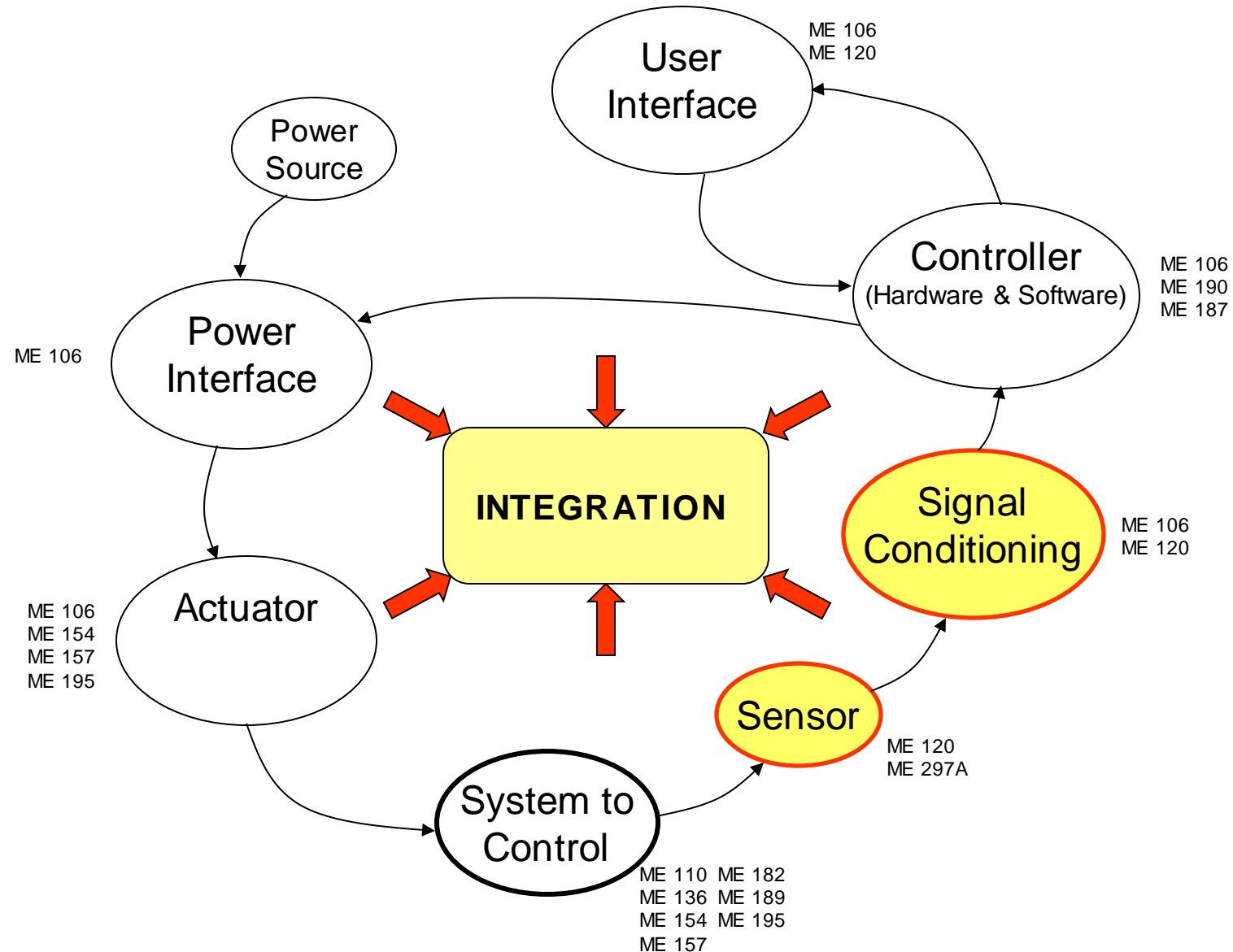
Ashby Building

Queen's University Belfast

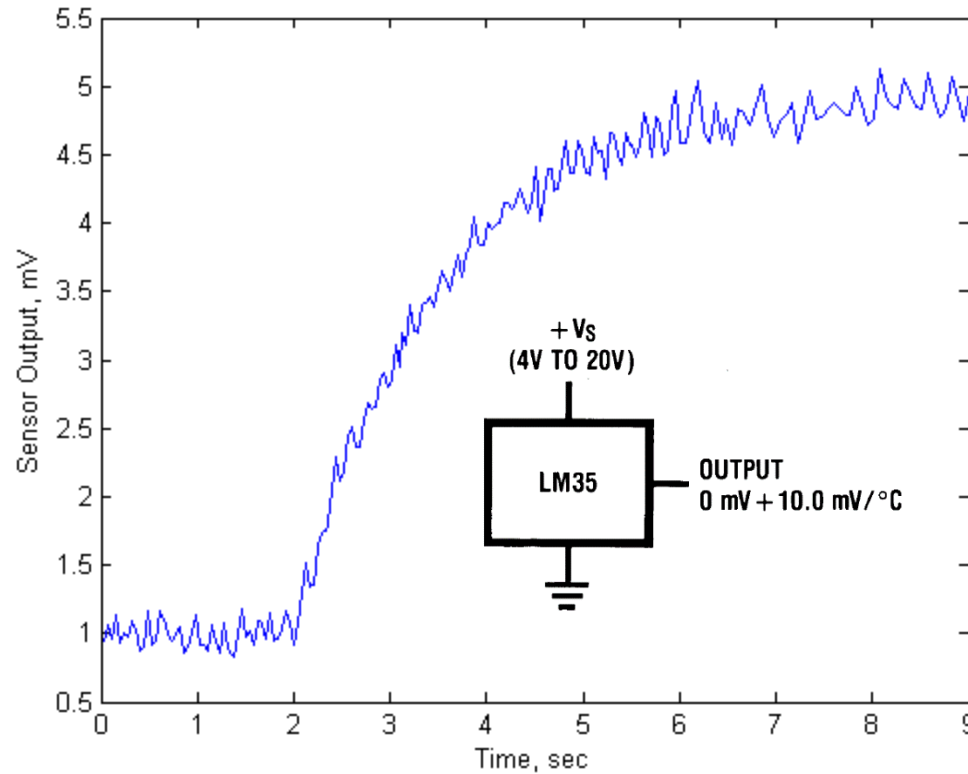
Aims

1. Filtering a noisy signal
2. High-pass and low-pass
3. Magnitude and phase angle

Mechatronics Concept Map



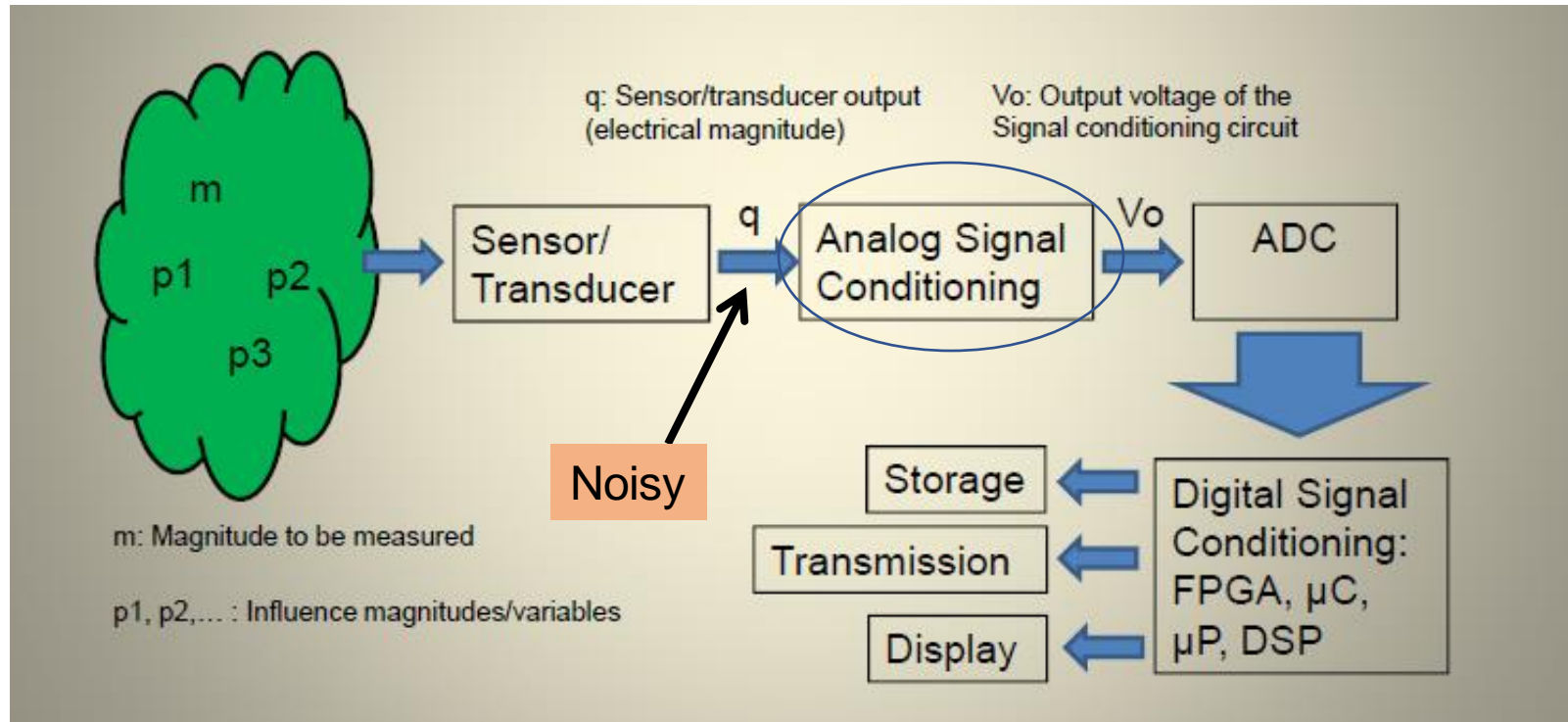
How to Handle Noisy Signals?



Filter!

Inset from: <http://cache.national.com/ds/LM/LM35.pdf>

Signal conditioning

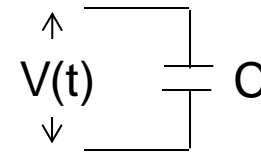


Impedance of a Capacitor

- Derive the impedance of a capacitor

$$Q(t) = CV(t) \quad (1) \quad \text{Physics for a capacitor}$$

$$\text{Let } V(t) = Ae^{st} \quad (2) \quad \text{A time varying function}$$



$$\frac{dQ(t)}{dt} = i(t) = C \frac{dV(t)}{dt} = CsAe^{st} \quad (3) \quad \text{Differentiate both sides}$$

$$\text{Define } Z(t) = \frac{V(t)}{i(t)} \quad (4) \quad \text{The impedance – the ratio of voltage to current}$$

$$\therefore Z(t) = \frac{Ae^{st}}{CsAe^{st}} = \frac{1}{Cs} \quad (5)$$

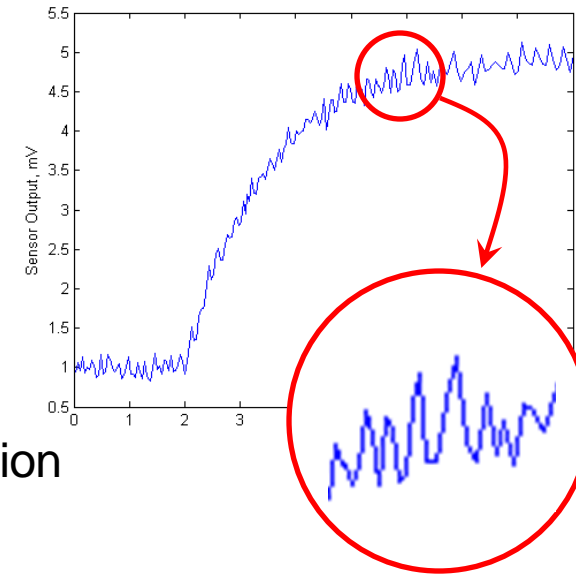
What is s?

What is s?

- Previously

$$V(t) = A e^{st} \quad (2)$$

A time varying function



$$\text{Let } s = j\omega \quad (6)$$

$$\text{where } j = \sqrt{-1}$$

and ω = circular frequency, in rad/s

$$\longrightarrow \therefore Z_C = \frac{1}{jC\omega}$$

■ Why?

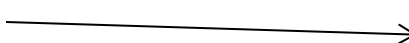
$$V(t) = Ae^{j\omega t} = A[\cos(\omega t) + j \sin(\omega t)]$$

(7) A sinusoidal function,
which can also be thought
of as rotating vector

Complex Numbers and Vectors

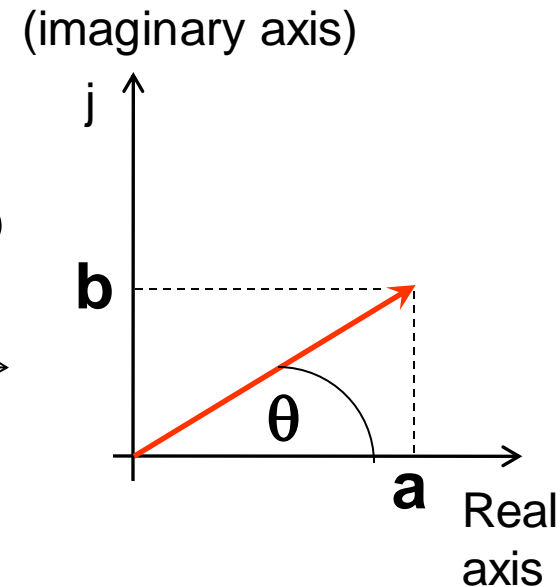
- Think of a complex number as a **vector**
 - Vectors have:
 - Magnitude (length)
 - Direction (angle)

$$V(t) = Ae^{j\omega t} = A[\cos(\omega t) + j \sin(\omega t)] \quad (7)$$

is of the form: $a + jb$ 

$$\text{Magnitude} = \sqrt{a^2 + b^2}$$

$$\text{Theta} = \text{Direction} = \tan^{-1}\left(\frac{b}{a}\right)$$

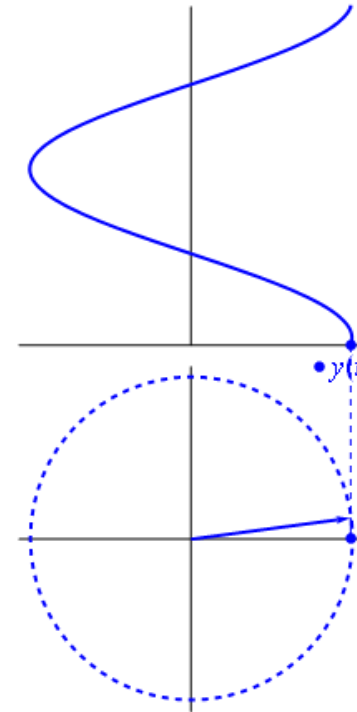


Sinusoidal Function

- Visualize the connection between the vector and the sinusoidal function of time
 - Suppose the real component is plotted as a function of time

$$V(t) = Ae^{j\omega t} = A[\cos(\omega t) + j \sin(\omega t)] \quad (7)$$

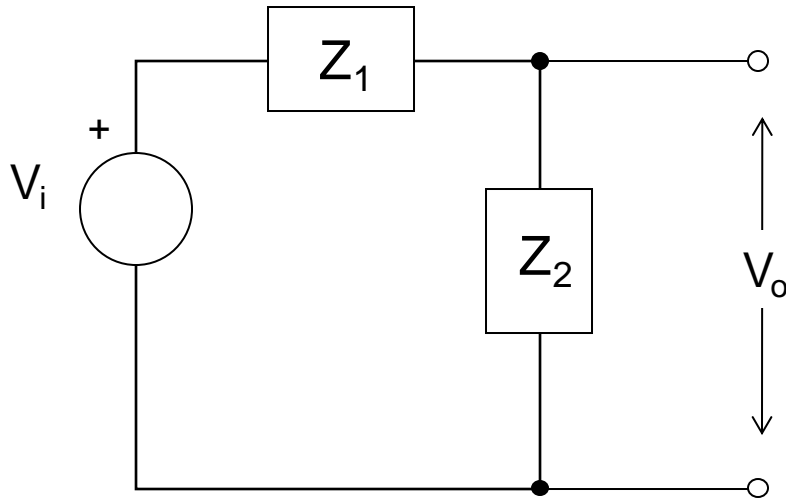
$$V_{real}(t) = A\cos(\omega t) \quad (8)$$



<http://upload.wikimedia.org/wikipedia/commons/8/89/Unfasor.gif>

Generalized Voltage Divider

- What is V_o in terms of V_i , Z_1 , and Z_2 ?



$$V_o = V_i \left[\frac{Z_2}{Z_1 + Z_2} \right] \quad (9)$$

$$\frac{V_o}{V_i} = \left[\frac{Z_2}{Z_1 + Z_2} \right]$$

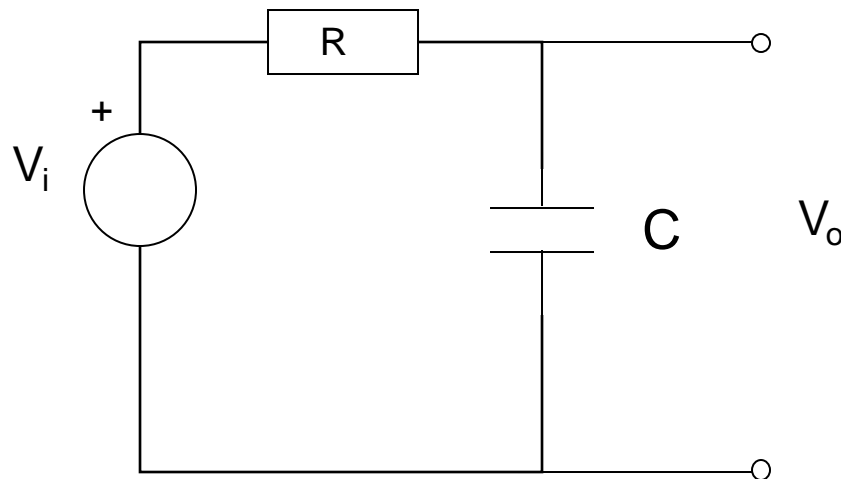
the 'transfer function'

RC Filters

- Frequency dependent voltage divider

- Impedance of a resistor, R
- Impedance of a capacitor, Z_C

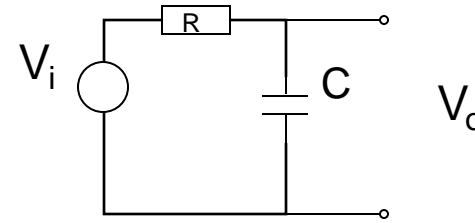
$$Z_C = \frac{1}{jC\omega} \quad j = \sqrt{-1}$$



$$\frac{V_o}{V_i} = \left[\frac{Z_2}{Z_1 + Z_2} \right]$$

$$\frac{V_o}{V_i} = \frac{\frac{1}{jC\omega}}{\frac{1}{jC\omega} + R} = \frac{1}{1 + jRC\omega}$$

RC Filters



- Transfer function
 - Complex number
 - Magnitude

$$\frac{V_o}{V_i} = \frac{1}{1 + jRC\omega}$$

- (Magnitude of numerator)/(Magnitude of denominator)

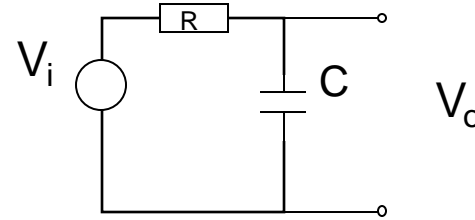
$$\left| \frac{V_o}{V_i} \right| = \frac{1}{\sqrt{1 + (RC\omega)^2}}$$

- Angle (“phase angle”)

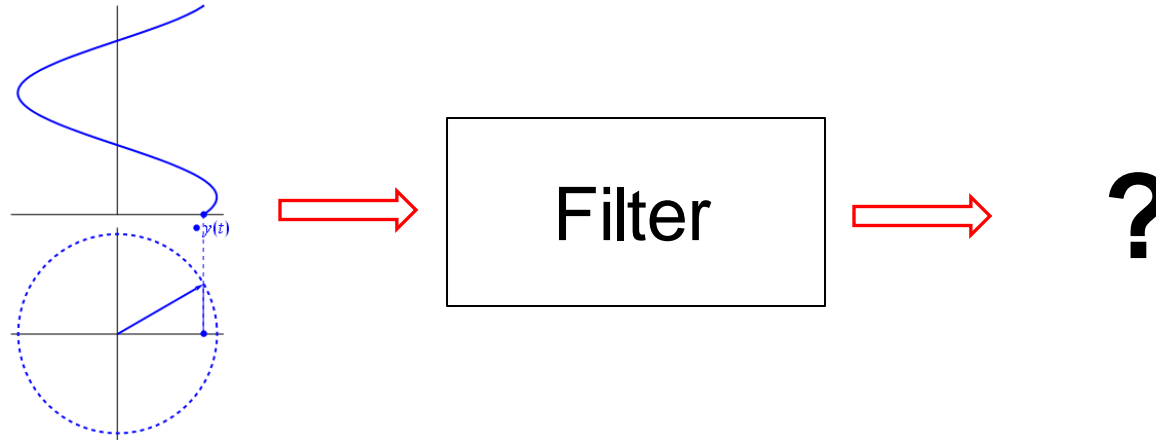
- (angle of num) – (angle of denom)
- How much the output is out of time synchronization with input

$$\angle \frac{V_o}{V_i} = -\tan^{-1}(RC\omega)$$

RC Filters



- Behavior
 - How does the magnitude and angle of the transfer function change *with frequency*?
 - Ex. $R=10k$, $C=1\mu F$

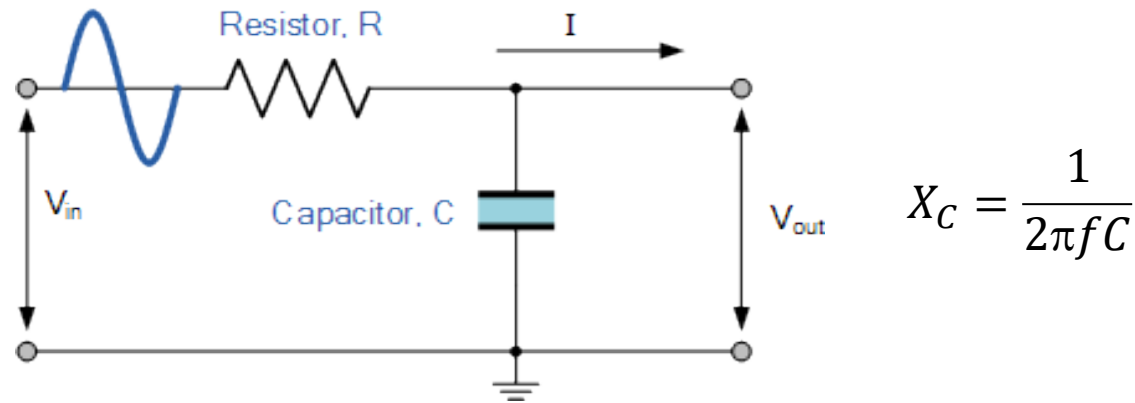


FILTERING

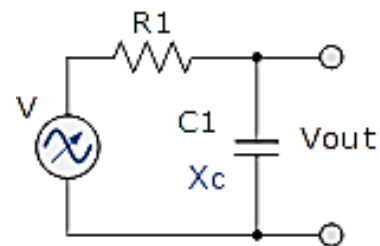
- The three most common types of filters are called
 - Butterworth
 - fairly flat response in the pass-band
 - a steep attenuation rate,
 - a non-linear phase response
 - Chebyshev
 - steeper rate of attenuation, develop some ripple in the pass band.
 - The phase response is much more non-linear than the Butterworth.
 - Bessel
 - have the best step response and phase linearity. **But requires a number of stages or orders**

LOW-PASS FILTER

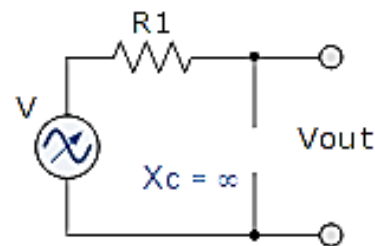
- A low-pass filter allows for easy passage of low-frequency signals from source to load, and difficult passage of high-frequency signals.
- The cutoff frequency for a low-pass filter is that frequency at which the output (load) voltage equals 70.7% of the input (source) voltage.
- Above the cutoff frequency, the output voltage is lower than 70.7% of the input, and vice versa.



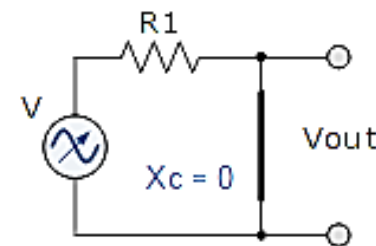
First order low pass filter



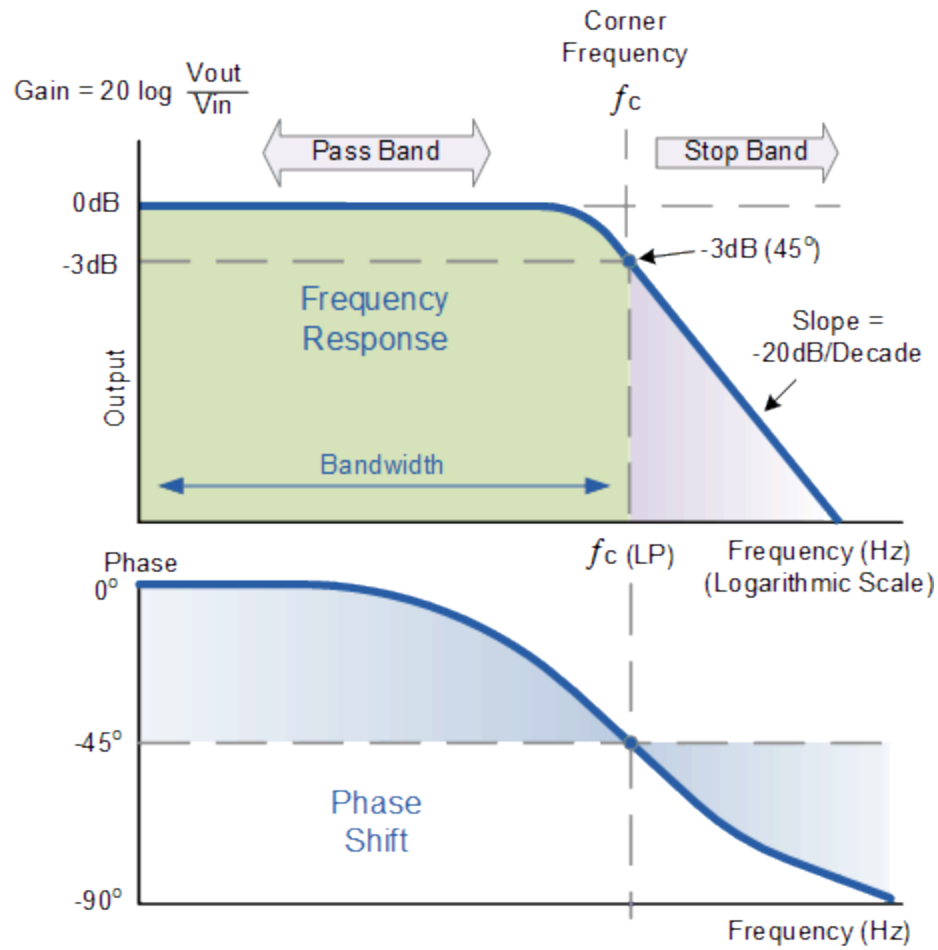
Low Pass at normal frequency



Low Pass at DC zero frequency



Low Pass at high frequency



Cut-off Frequency and Phase Shift

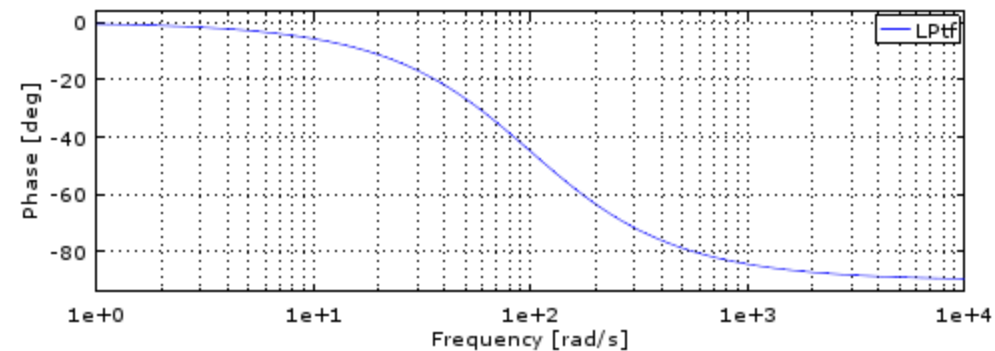
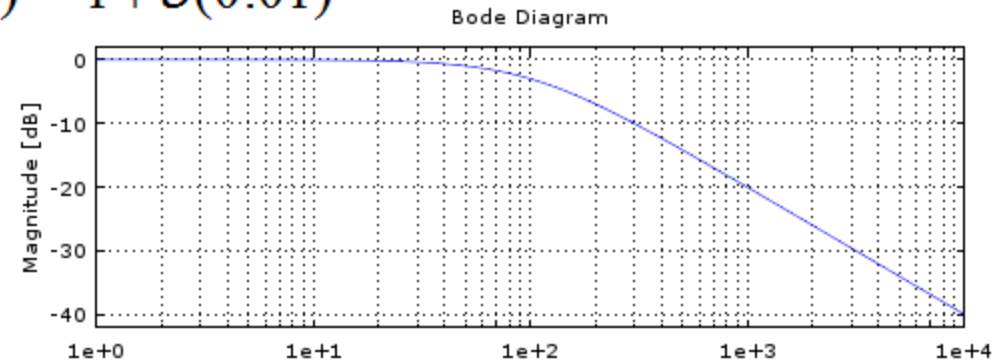
$$f_c = \frac{1}{2\pi RC}$$

$$\text{Phase Shift } \phi = -\arctan(2\pi fRC)$$

Bode Plots – Matlab/Octave Style

$$\begin{aligned}\frac{V_o}{V_i} &= \frac{1}{1 + j\omega RC} = \frac{1}{1 + sRC} \\ &= \frac{1}{1 + s(1 \times 10^4 \Omega)(1 \times 10^{-6} F)} = \frac{1}{1 + s(0.01)}\end{aligned}$$

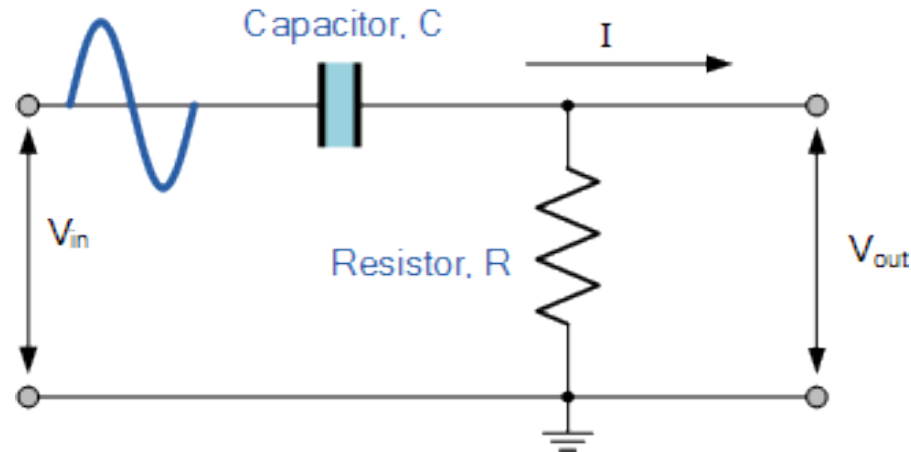
```
>> s = tf('s');  
>> LPtf = 1/(1+s*0.01);  
>> bode(LPtf)
```



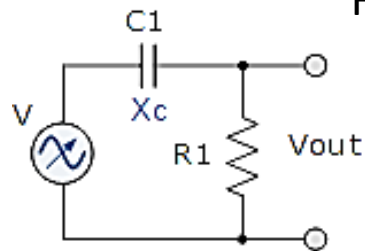
HIGH-PASS FILTER

A high-pass filter allows for easy passage of high-frequency signals from source to load, and difficult passage of low-frequency signals.

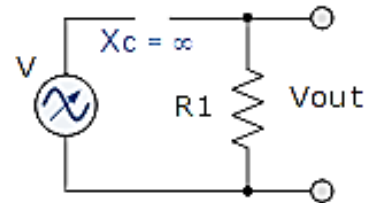
The cutoff frequency for a high-pass filter is that frequency at which the output (load) voltage equals 70.7% of the input (source) voltage. Above the cutoff frequency, the output voltage is greater than 70.7% of the input, and vice versa.



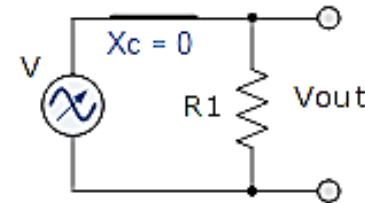
First order high pass filter



High Pass at normal frequency

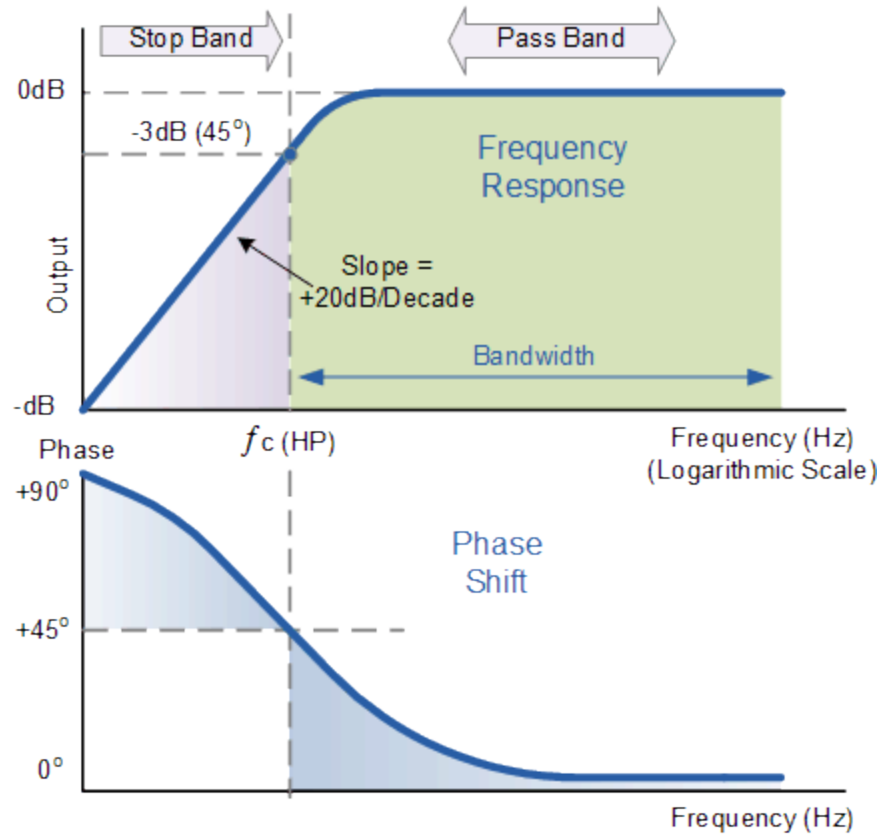


High Pass at DC zero frequency



High Pass at high frequency

$$\text{Gain (dB)} = 20 \log \frac{V_{\text{out}}}{V_{\text{in}}}$$



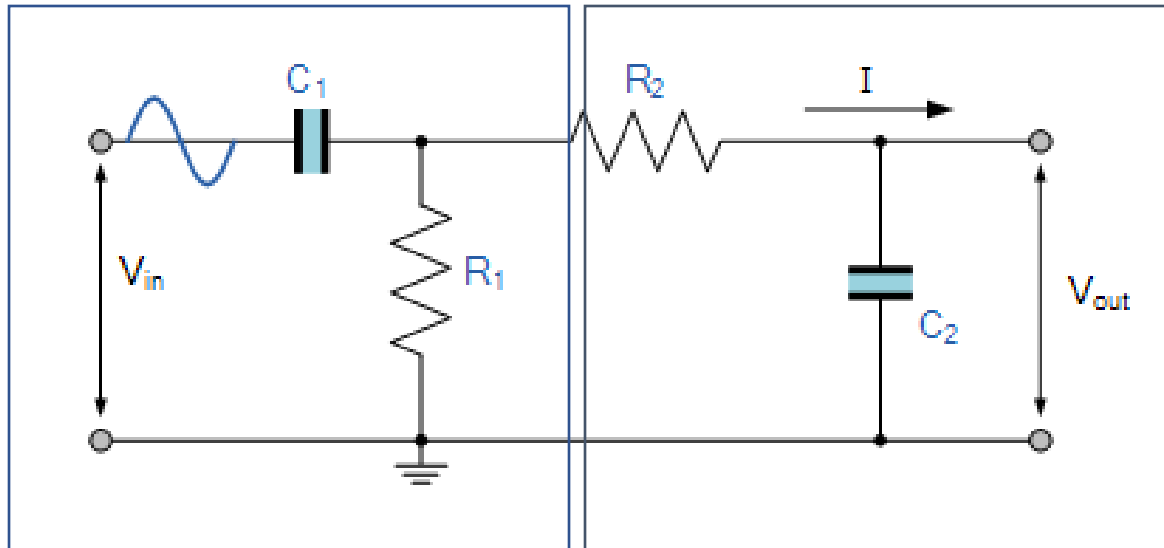
Cut-off Frequency and Phase Shift

$$f_c = \frac{1}{2\pi RC}$$

$$\text{Phase Shift } \phi = -\arctan(2\pi fRC)$$

BAND-PASS FILTER

- By connecting or “cascading” together a single High Pass Filter circuit with a Low Pass Filter circuit, we can produce another type of passive RC filter that passes a selected range or “band” of frequencies that can be either narrow or wide while attenuating all those outside of this range.
- known commonly as a Band Pass Filter



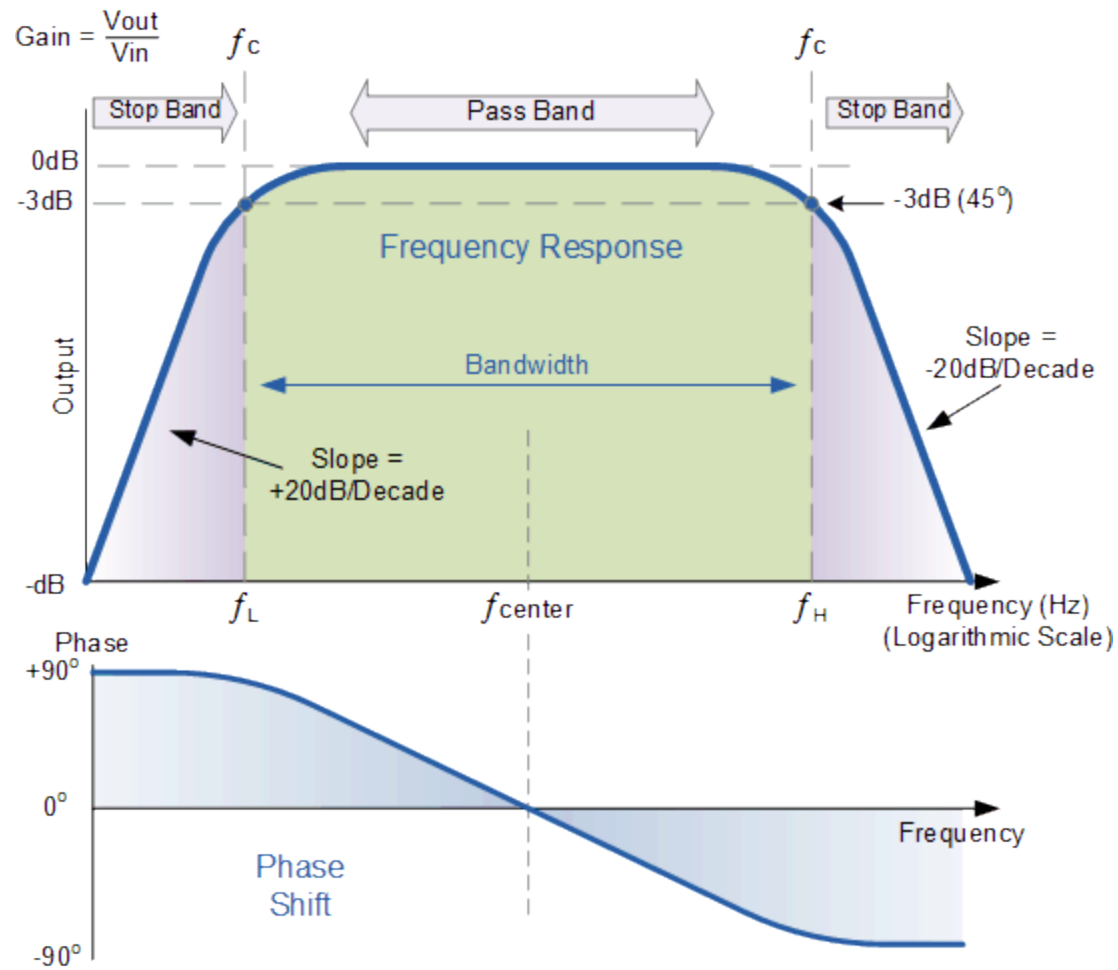
To set the first band
pass frequency (HP)

To set the second band
pass frequency (LP)

Cut-off Frequency and Phase Shift

$$f_c = \frac{1}{2\pi RC}$$

Phase Shift $\phi = -\arctan(2\pi fRC)$



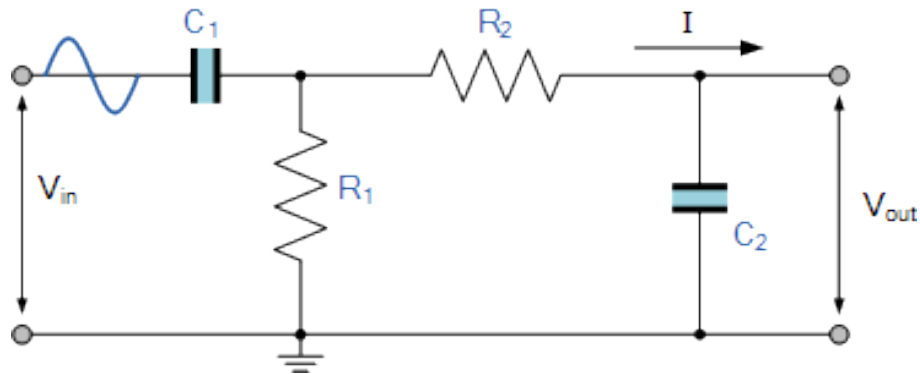
Cut-off Frequency and Phase Shift

$$f_c = \frac{1}{2\pi RC}$$

Phase Shift $\phi = -\arctan(2\pi fRC)$

EXAMPLE 1

A second-order band pass filter is to be constructed using RC components that will only allow a range of frequencies to pass above 1kHz (1,000Hz) and below 30kHz (30,000Hz). Assuming that both the resistors have values of $10\text{k}\Omega$'s, calculate the values of the two capacitors required



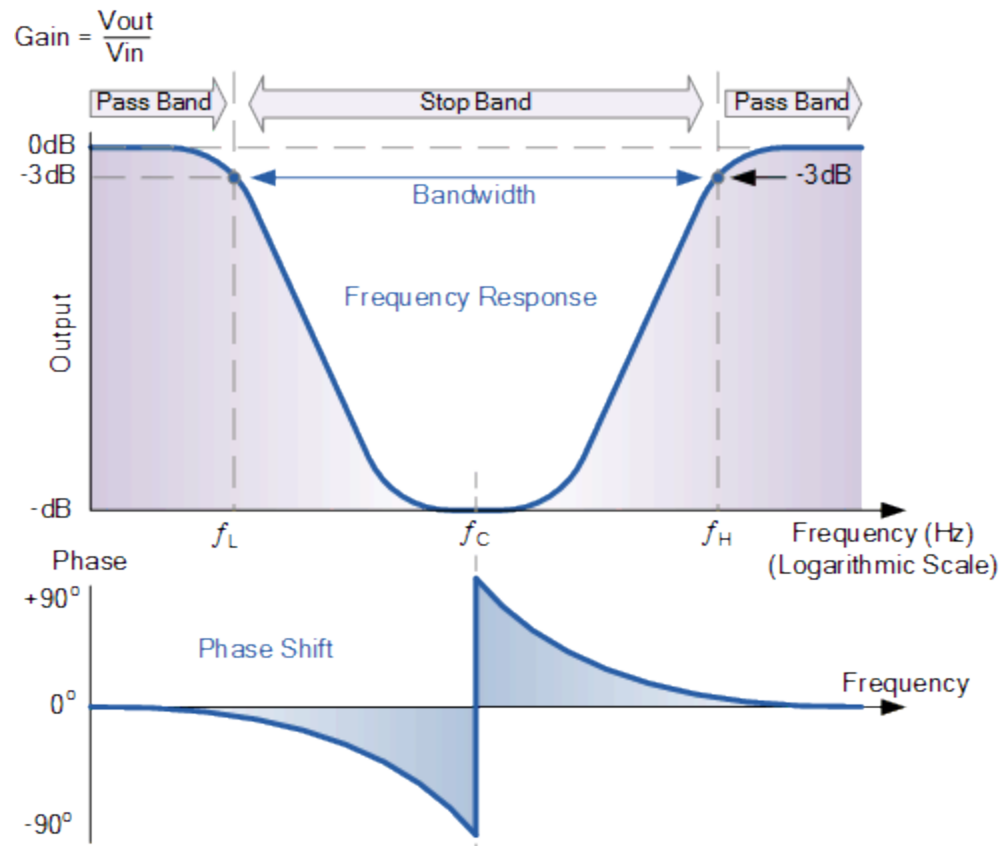
Answers:

$$C_1 = 530 \text{ pF}$$

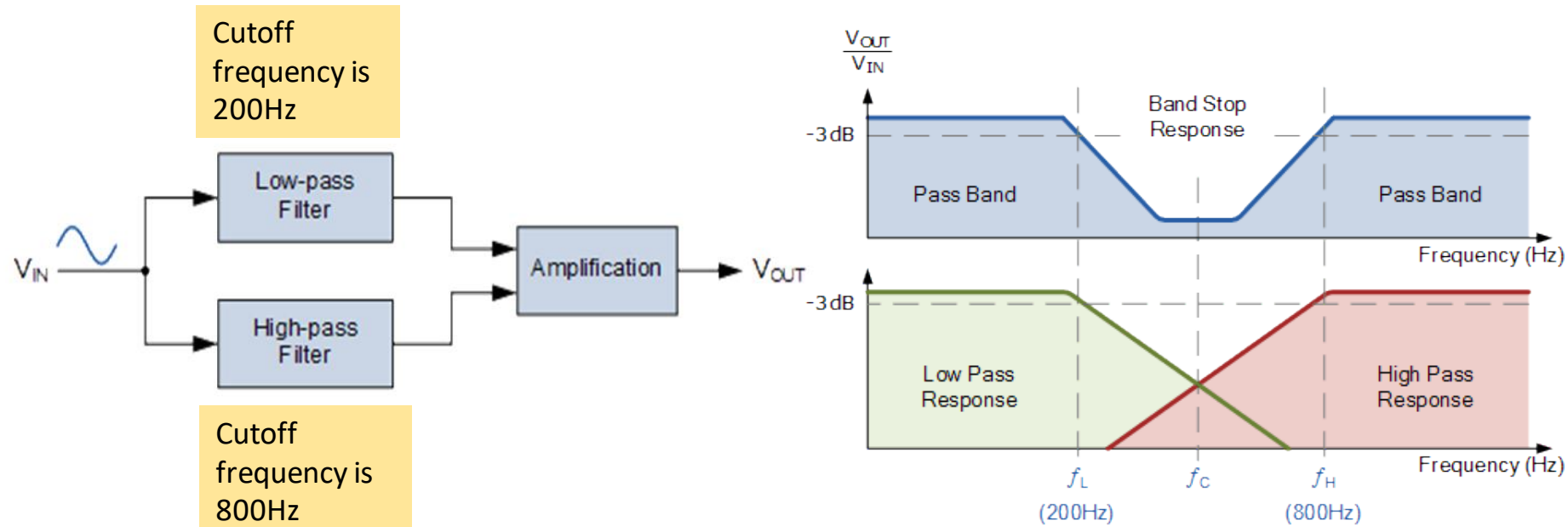
$$C_2 = 15.8 \text{ nF}$$

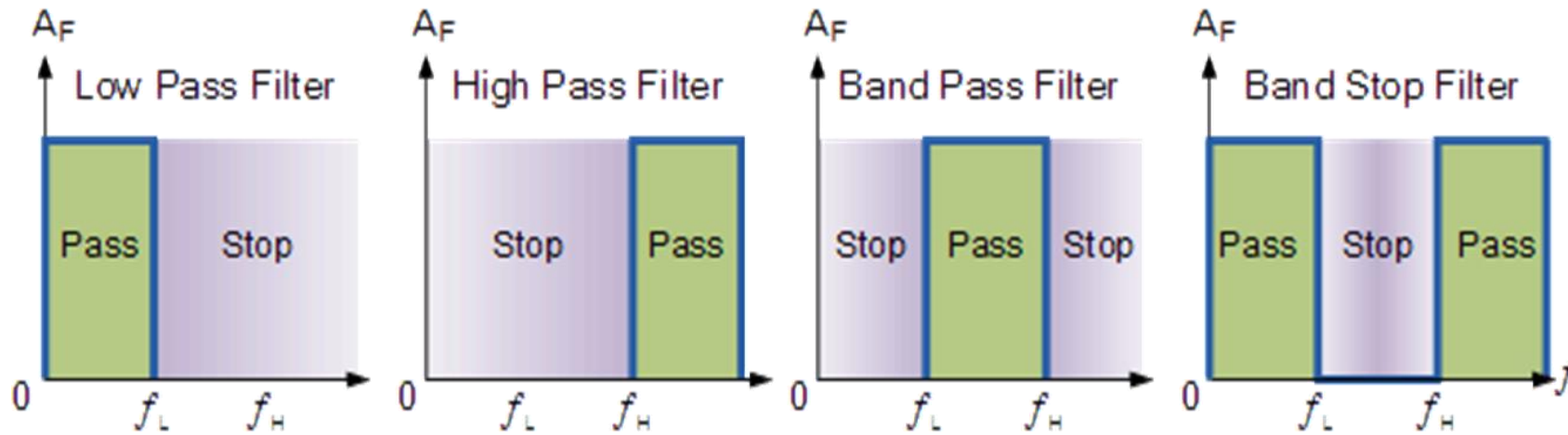
BAND-STOP FILTER

- combine the low and high pass filter to produce another kind of RC filter network
- that can block or at least severely attenuate a band of frequencies within these two cut-off frequency points.



- Band-pass filters are constructed by combining a low pass filter in series with a high pass filter
- Band stop filters are created by combining together the low pass and high pass filter sections in a “parallel” type configuration as shown.





Passband ?

Gain : dB

dB Hz

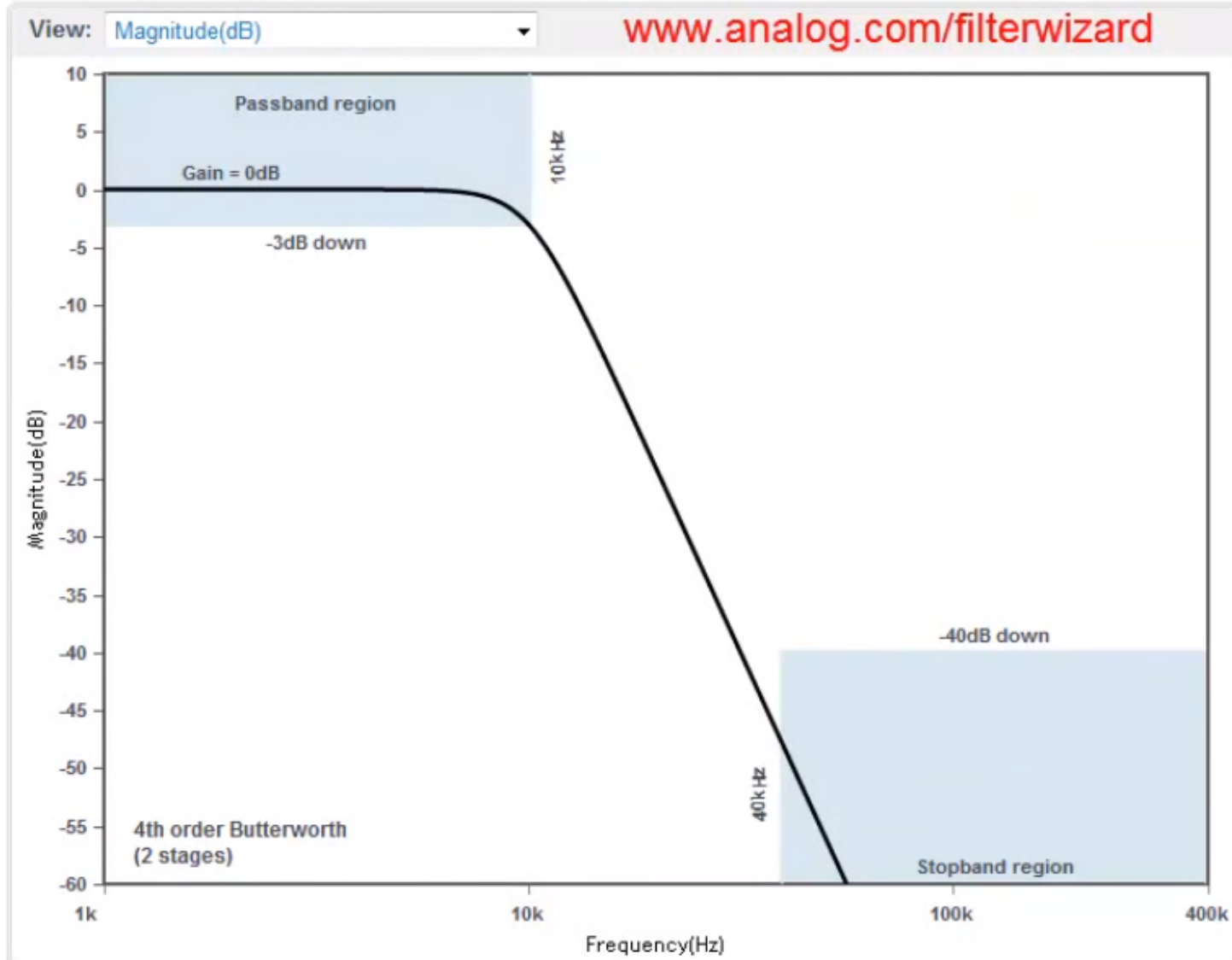
Stopband ?

dB Hz

Filter Response ?

Fewest Stages **Butterworth** Fastest Settling

*4th order Butterworth
(2 stages)*



Passband ?

Gain : dB

dB Hz

Stopband ?

dB Hz

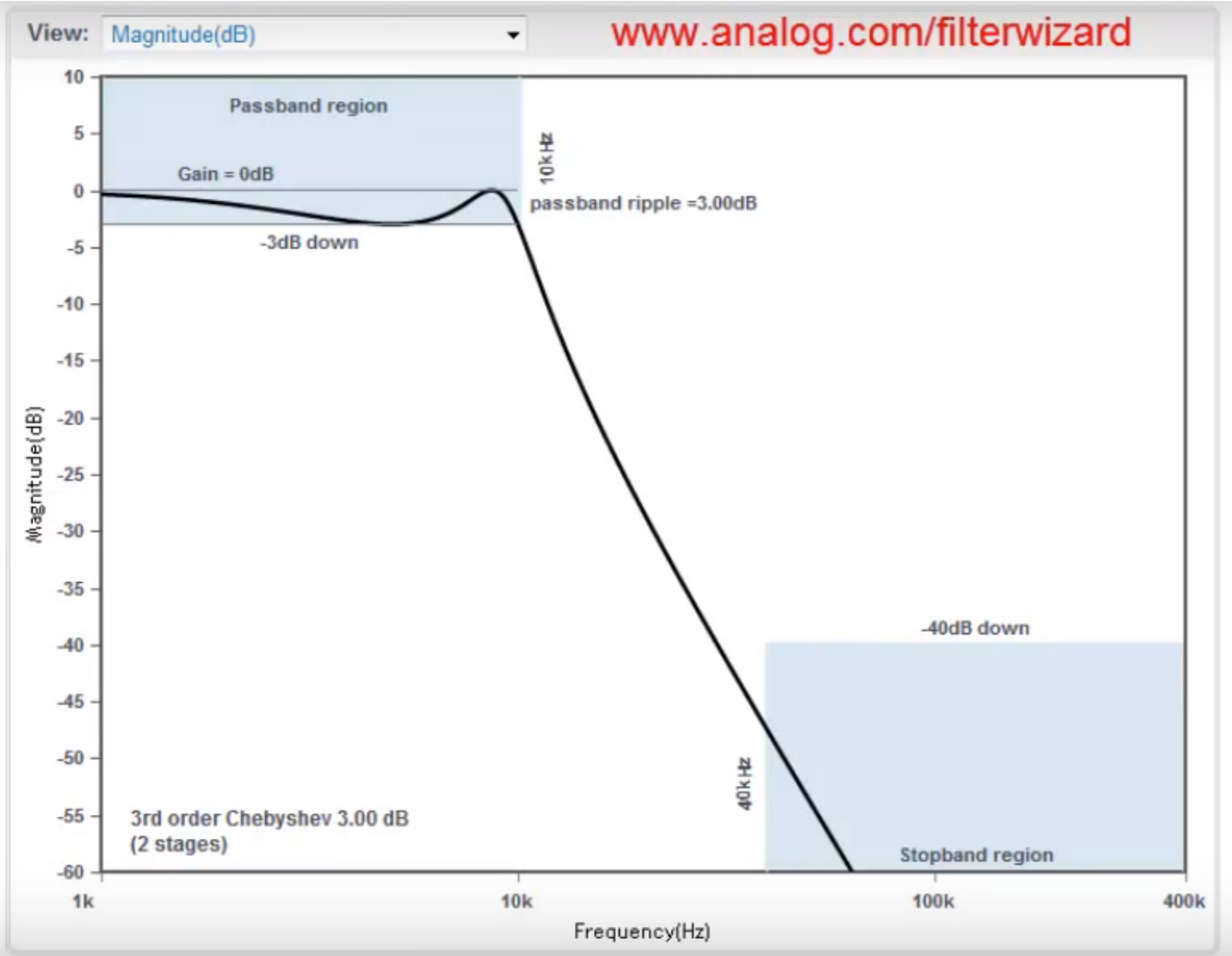
Filter Response ?

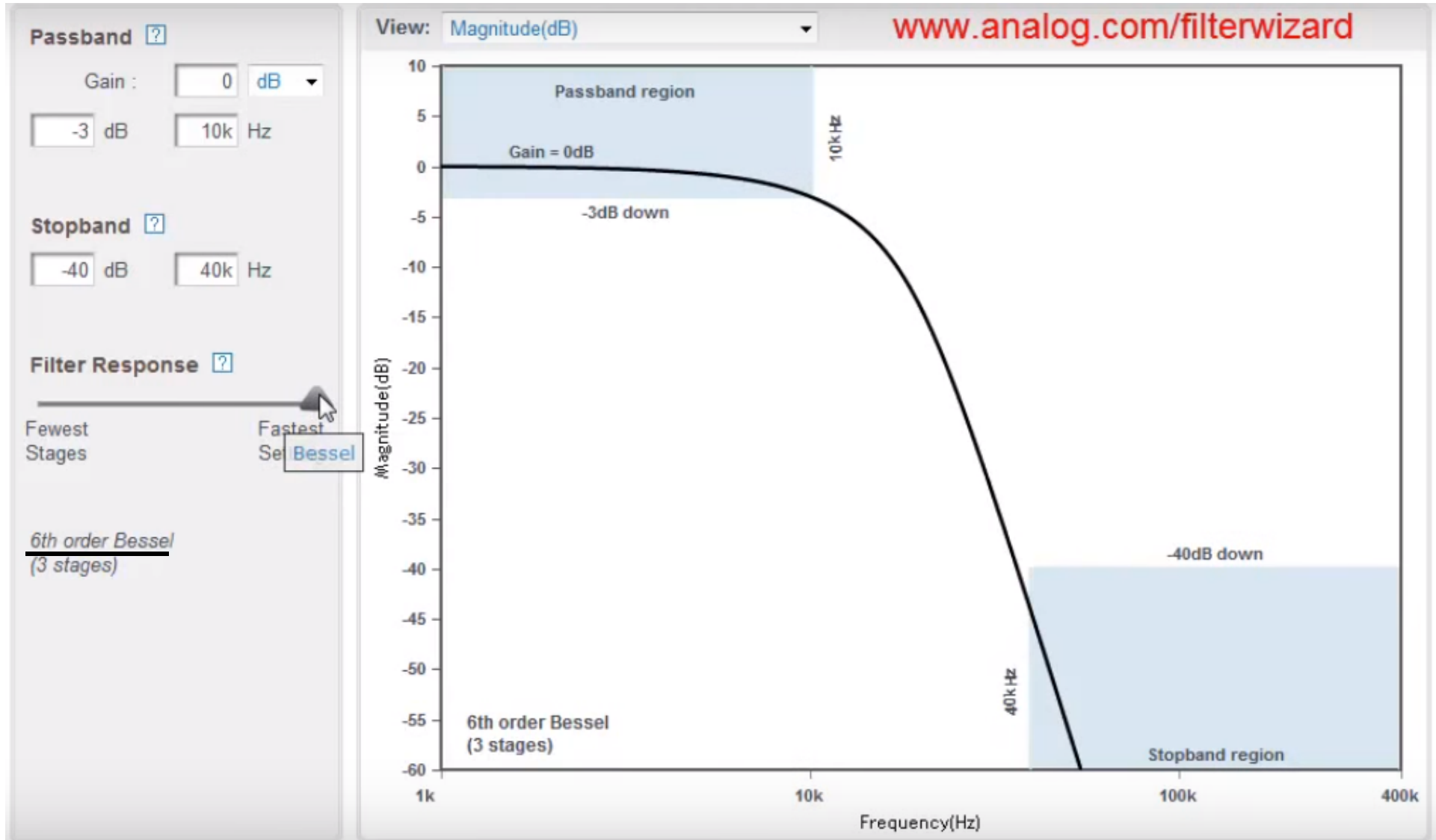
Fewest Settling

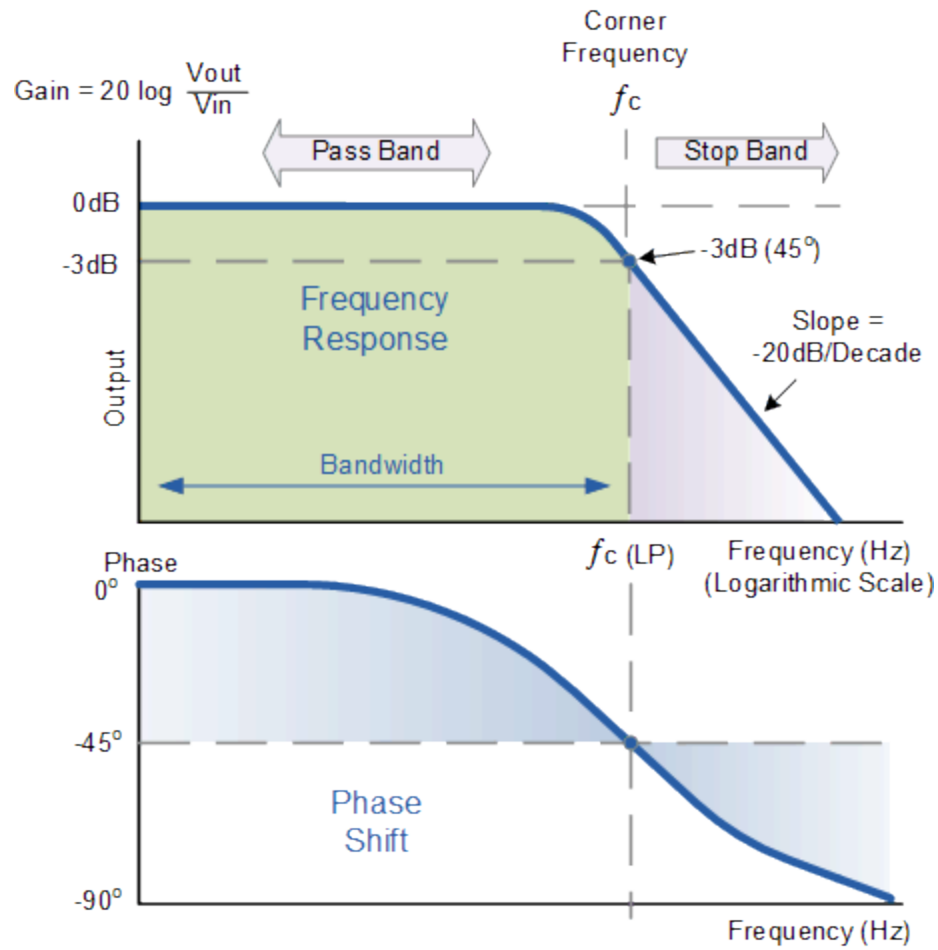
Chebyshev 3.00dB

Fastest Settling

3rd order Chebyshev 3.00 dB
(2 stages)







At low frequency (Pass Band)

$$\frac{V_{out}}{V_i} = Gain, A_v$$

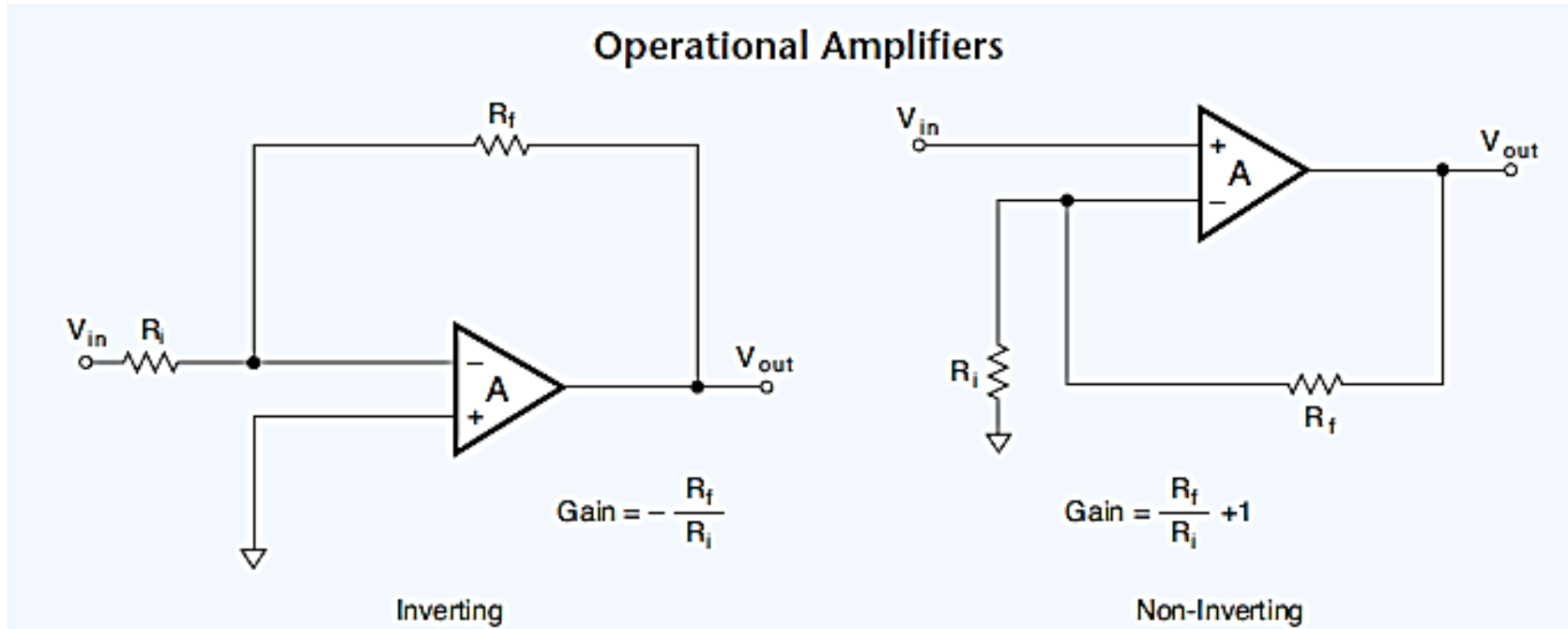
At cutoff frequency

$$\frac{V_{out}}{V_i} = 0.7071 A_v$$

AMPLIFICATION

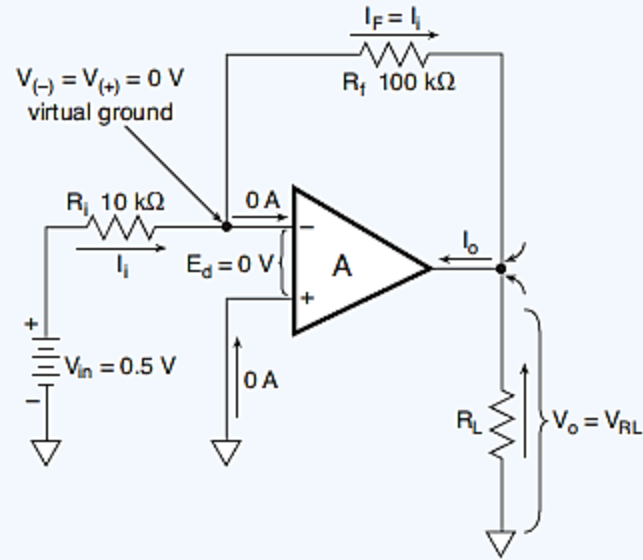
- Many sensors develop extremely low-level output signals
- Two common examples of low-level sensors are thermocouples and strain-gauge (less than 50 mV)
- Requires amplification
 - Using Operational Amplifier (Op-Amp)
 - Inverting Amplifier
 - Non-Inverting Amplifier
 - Differential Amplifier
 - Instrumentation Amplifier

INVERTING AND NON-INVERTING AMPLIFIERS



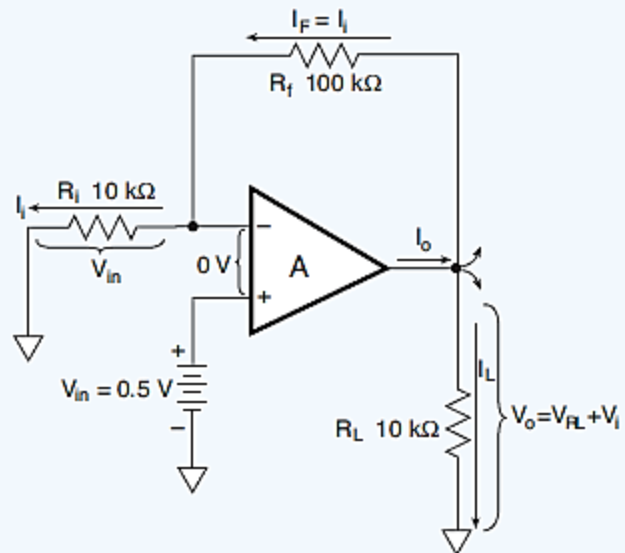
NOTE: The maximum input signal that the amplifier can handle without damage is usually about 2 V less than the supply voltage. For example, when the supply is $\pm 15 V_{DC}$, the input signal should not exceed $\pm 13 V_{DC}$.

Inverting Amplifier Stage



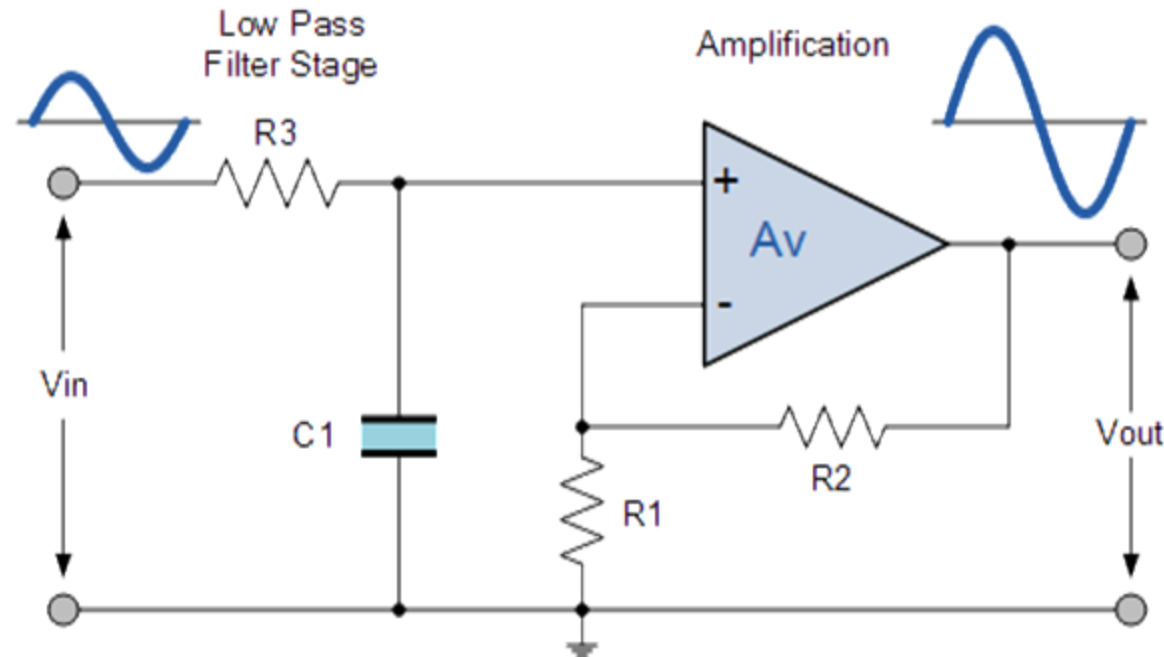
Gain = -10
 $V_o = -5 \text{ V}$

Non-Inverting Amplifier Stage



Gain = 11
 $V_o = 5.5 \text{ V}$

- Passive filters use R, L and C while Active Filters use R, L and C with combination of an active component such as Op-Amp.
- Main disadvantage of passive filters is that the amplitude of the output signal is less than that of the input signal

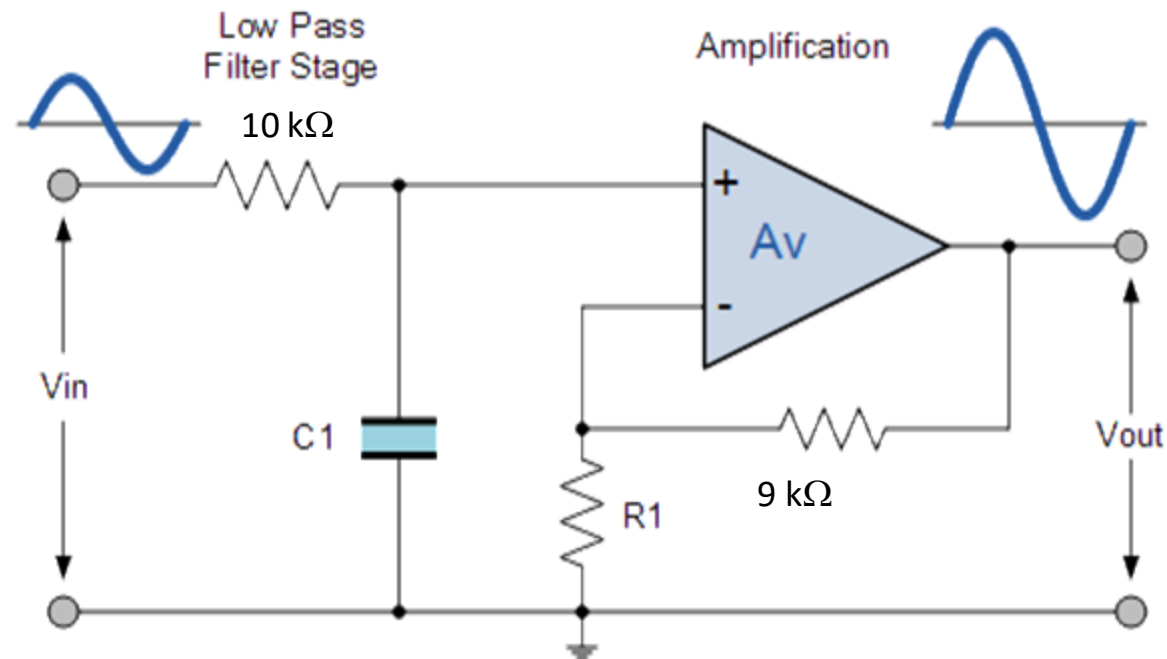


at low f : $X_c \rightarrow \infty$, $V_{out} = V_{in}$
 at high f : $X_c \rightarrow 0$, $V_{out} = 0$

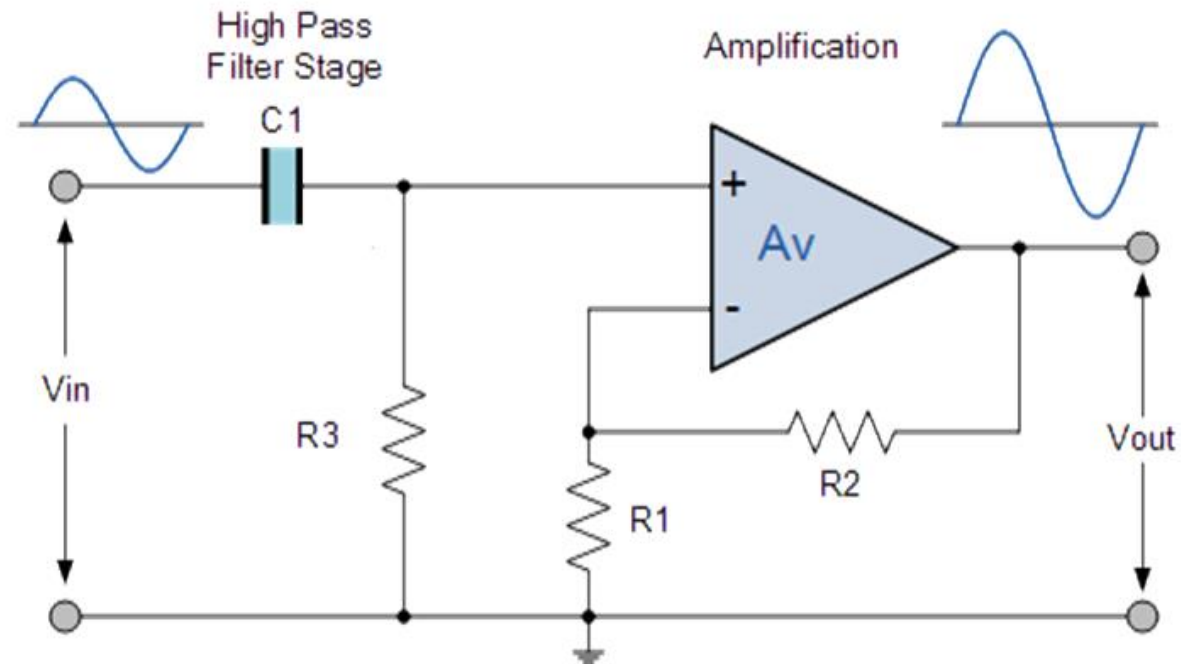
NON - INVERTING
 CONFIGURATION

$$\text{Gain, } A_v = 1 + \frac{R_2}{R_1}$$

- Design a non-inverting active low pass filter circuit that has a gain of ten at low frequencies, given that the resistor of the filter is $10\text{ k}\Omega$, the feedback resistor is $9\text{ k}\Omega$ and a high frequency cut-off or corner frequency of 159 Hz



$$R_1 = 1\text{ k}\Omega$$
$$C_1 = 100\text{ nF}$$

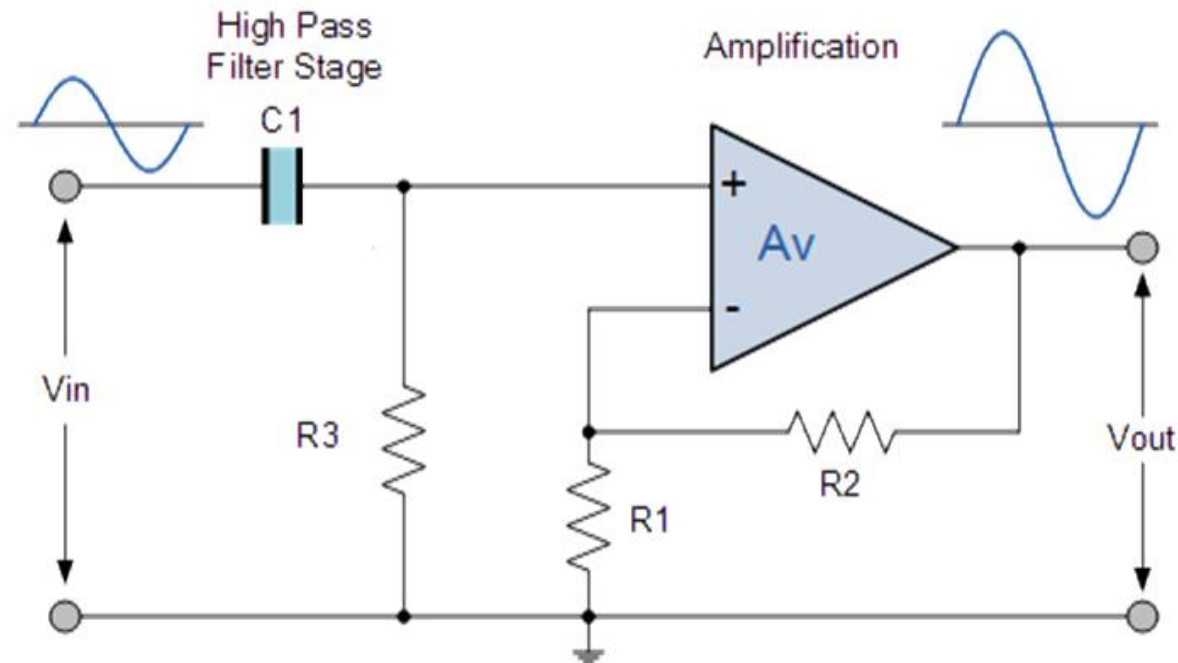


NON - INVERTING
CONFIGURATION

at low f : $X_c \rightarrow \infty$, $V_{out} = 0$
 at high f : $X_c \rightarrow 0$, $V_{out} = V_{in}$

$$\text{Gain, } A_v = 1 + \frac{R_2}{R_1}$$

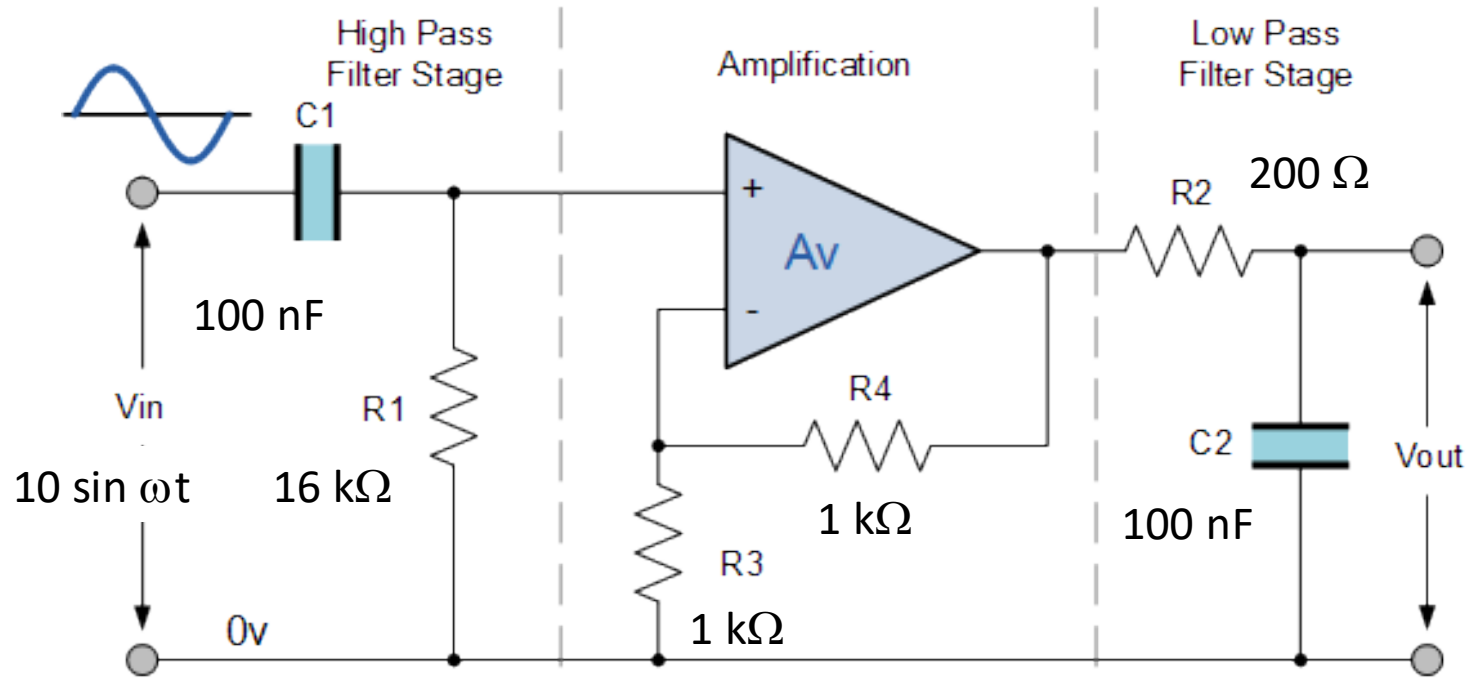
A first order active high pass filter has a pass band gain of two and a cut-off corner frequency of 1 kHz. If the input capacitor has a value of 10 nF, calculate the value of the resistor of the filter and the resistors of the amplifier.



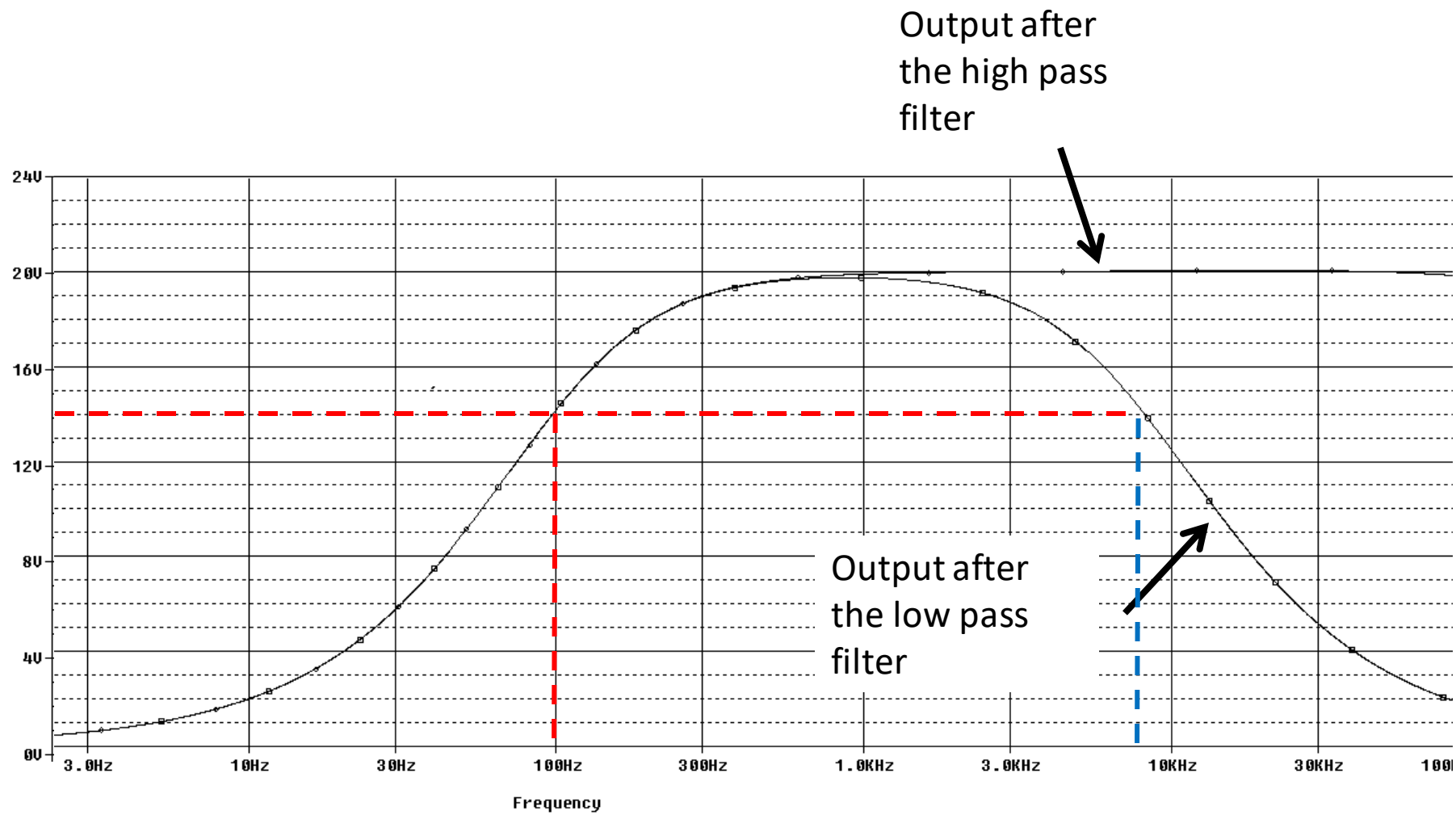
$$R_3 = 16 \text{ k}\Omega$$

$$R_1 = R_2$$

BAND PASS - ACTIVE FILTER



NON - INVERTING
CONFIGURATION



Thank You For Your Attention!

Any Question?

